

DFT Analysis (5B)

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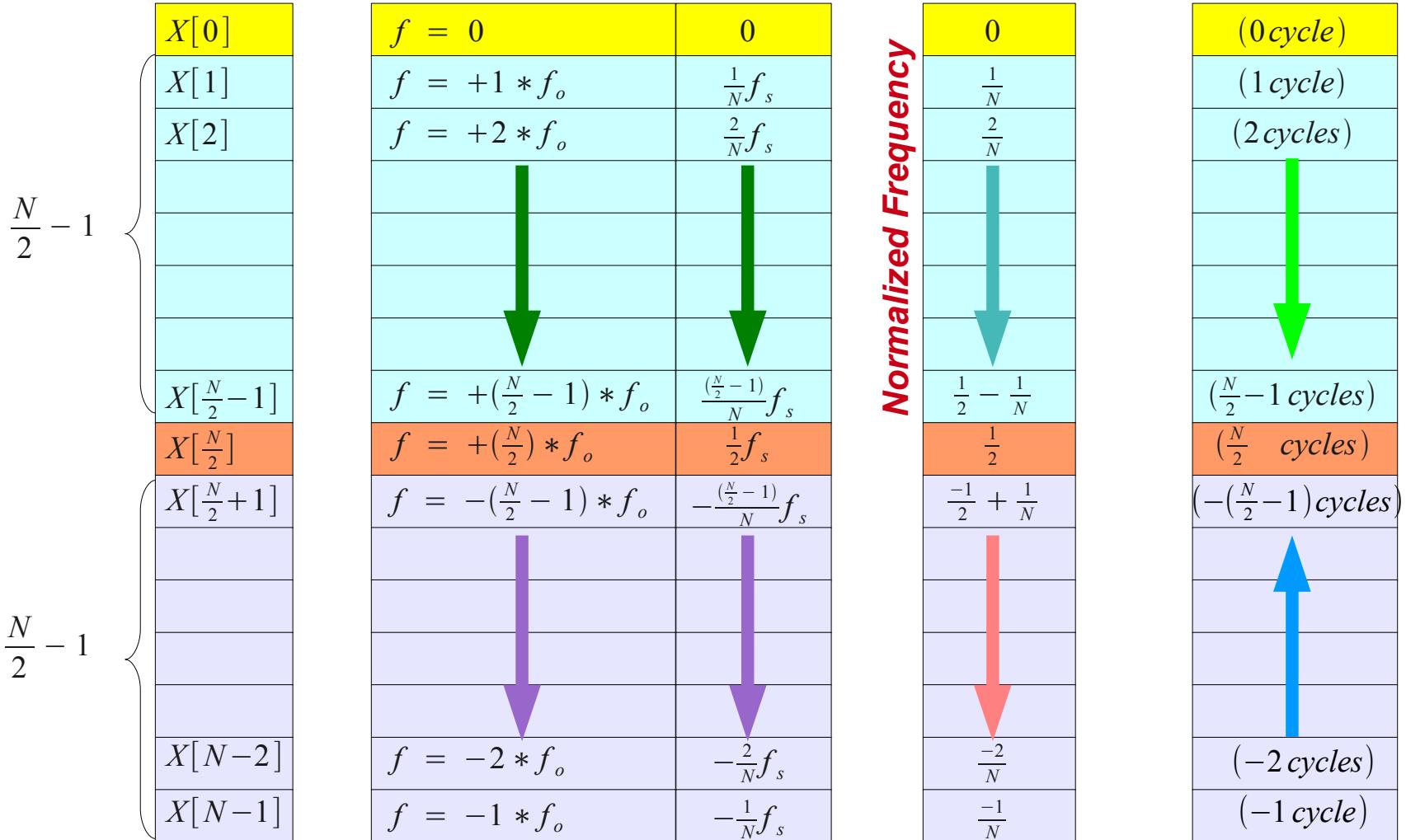
Frequency View of a DFT Matrix

<i>row 0</i>	$f = 0$	(0 cycle)	0
<i>row 1</i>	$f = -1 * f_o$	(-1 cycle)	$\frac{-1}{N}$
<i>row 2</i>	$f = -2 * f_o$	(-2 cycles)	$\frac{-2}{N}$
$\left\{ \begin{array}{l} \text{row } (\frac{N}{2}-1) \\ \text{row } (\frac{N}{2}) \end{array} \right.$	$f = -(\frac{N}{2}-1) * f_o$	$(-(\frac{N}{2}-1) \text{ cycles})$	$\frac{-1}{2} + \frac{1}{N}$
	$f = -(\frac{N}{2}) * f_o$	$(\frac{N}{2} \text{ cycles})$	$\frac{-1}{2}$
$\left\{ \begin{array}{l} \text{row } (\frac{N}{2}+1) \\ \vdots \\ \text{row } N-2 \\ \text{row } N-1 \end{array} \right.$	$f = +(\frac{N}{2}-1) * f_o$	$(\frac{N}{2}-1 \text{ cycles})$	$\frac{1}{2} - \frac{1}{N}$
	$f = +2 * f_o$	(2 cycles)	$\frac{2}{N}$
	$f = +1 * f_o$	(1 cycle)	$\frac{1}{N}$

Normalized Frequency

$$f_o = \frac{f_s}{N}$$

Frequency View of a X[i] Vector



Frequency and Time Interval (1)

Freq Domain

$$\Delta f_1 = \frac{f_{s1}}{4} = \frac{1}{4\tau}$$
$$f_{h1} = \frac{f_{s1}}{2} = \frac{1}{2\tau}$$

$$\Delta f_2 = \frac{f_{s2}}{8} = \frac{1}{4\tau}$$
$$f_{h2} = \frac{f_{s2}}{2} = \frac{1}{\tau}$$

Time Domain

4-pt DFT

8-pt DFT

$\Delta t_1 = \tau$ $f_{s1} = \frac{1}{\tau}$

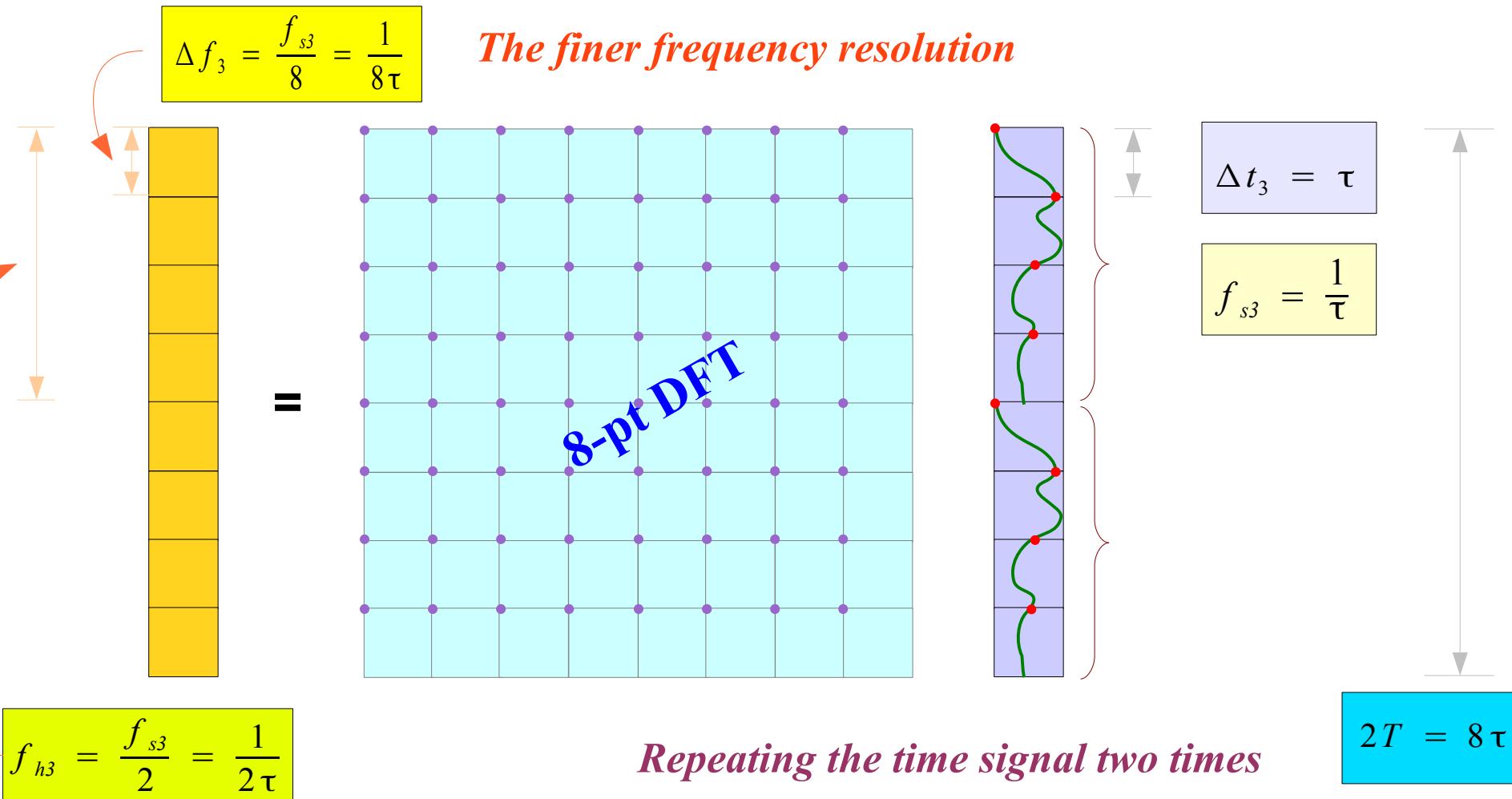
$\Delta t_2 = \tau/2$ $f_{s2} = \frac{2}{\tau}$

$T = 4\tau$

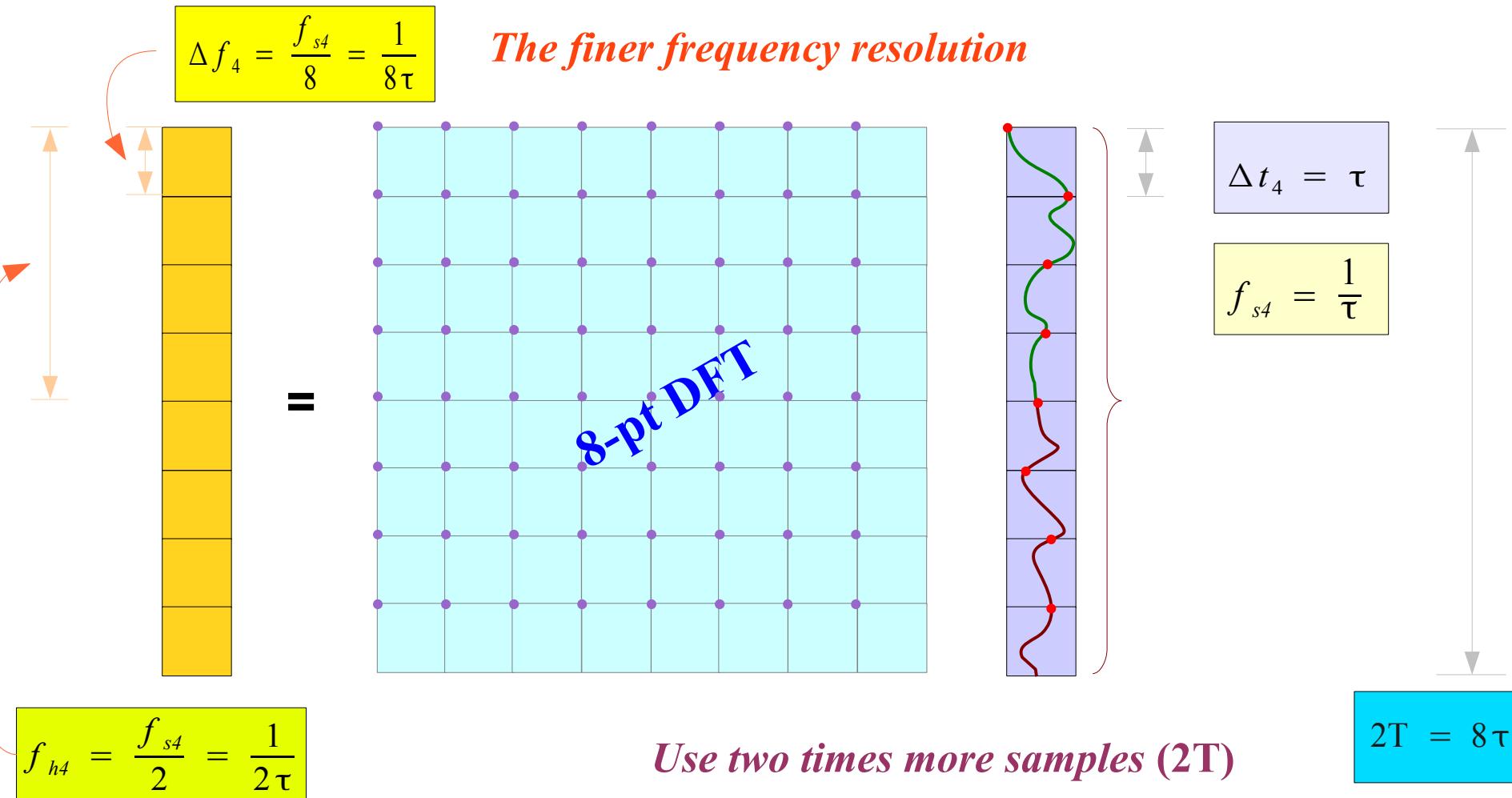
$T = 4\tau$

The same frequency resolution

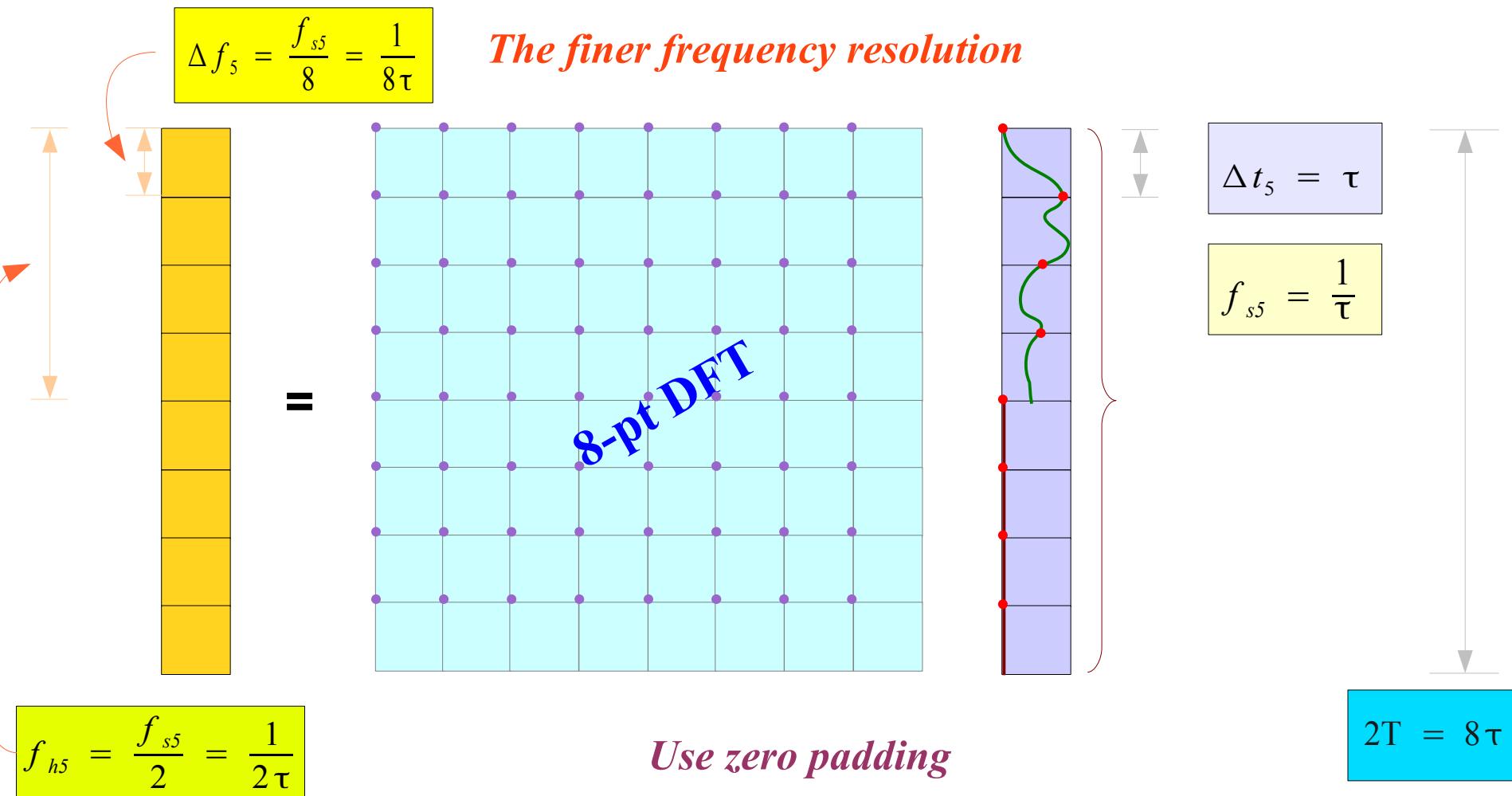
Frequency and Time Interval (2)



Frequency and Time Interval (3)



Frequency and Time Interval (4)



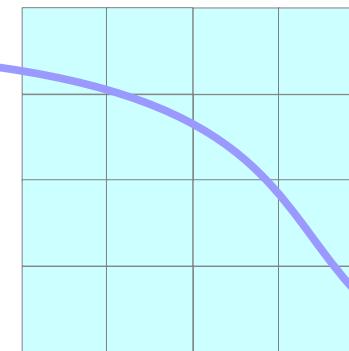
Frequency and Time Interval (5)

Freq Domain

$$f_h = \frac{f_s}{2} = \frac{1}{2\Delta t}$$

$$\Delta f = \frac{1}{T}$$

=



Time Domain

$$f_s = \frac{1}{\Delta t}$$

$$\Delta t$$

A vertical double-headed arrow indicating the relationship between the two domains, labeled T at the bottom.

$$T$$

Single-Sided Spectrum

$$x(t) = a_0 + \sum_{k=1}^{\infty} (a_k \cos(k\omega_0 t) + b_k \sin(k\omega_0 t))$$

$$a_0 = \frac{1}{T} \int_0^T x(t) dt$$

$$a_k = \frac{2}{T} \int_0^T x(t) \cos(k\omega_0 t) dt$$

$$b_k = \frac{2}{T} \int_0^T x(t) \sin(k\omega_0 t) dt$$

$$k = +1, +2, \dots$$

$$x(t) = g_0 + \sum_{k=1}^{\infty} g_k \cos(k\omega_0 t + \phi_k)$$

$$g_0 = a_0$$

$$g_k = \sqrt{a_k^2 + b_k^2}$$

$$\phi_k = \tan^{-1} \left(-\frac{b_k}{a_k} \right)$$

$$k = +1, +2, \dots$$

$$\cos(\alpha + \beta) = \underline{\cos(\alpha)} \cos(\beta) - \underline{\sin(\alpha)} \sin(\beta)$$

$$g_k \cos(k\omega_0 t + \phi_k) = \underline{g_k \cos(\phi_k)} \cos(k\omega_0 t) - \underline{g_k \sin(\phi_k)} \sin(k\omega_0 t)$$

$$\underline{a_k \cos(k\omega_0 t)} + \underline{b_k \sin(k\omega_0 t)}$$

$$a_k = g_k \cos(\phi_k)$$

$$-b_k = g_k \sin(\phi_k)$$

$$a_k^2 + b_k^2 = g_k^2$$

$$-\frac{b_k}{a_k} = \tan(\phi_k)$$

Two-Sided Spectrum

$$x(t) = \sum_{k=-\infty}^{+\infty} C_k e^{+jk\omega_0 t}$$

$$x(t) = \sum_{k=-\infty}^{+\infty} C_k e^{+jk\omega_0 t}$$

$$C_k = \frac{1}{T} \int_0^T x(t) e^{-jk\omega_0 t} dt$$

$$k = \dots, -2, -1, 0, +1, +2, \dots$$

$$C_k = \begin{cases} a_0 & (k = 0) \\ \frac{1}{2}(a_k - jb_k) & (k > 0) \\ \frac{1}{2}(a_k + jb_k) & (k < 0) \end{cases}$$

$$C_k = \frac{1}{T} \int_0^T x(t) e^{-jk\omega_0 t} dt$$

$$k = \dots, -2, -1, 0, +1, +2, \dots$$

$$|C_k| = \begin{cases} a_0 & (k = 0) \\ \frac{1}{2}\sqrt{a_k^2 + b_k^2} & (k \neq 0) \end{cases}$$

$$\text{Arg}(C_k) = \begin{cases} \tan^{-1}(-b_k/a_k) & (k > 0) \\ \tan^{-1}(+b_k/a_k) & (k < 0) \end{cases}$$

$$C_k = \begin{cases} a_0 & (k = 0) \\ \frac{1}{2}g_k e^{+jk\phi_k} & (k > 0) \\ \frac{1}{2}g_k e^{-jk\phi_k} & (k < 0) \end{cases}$$

$$|C_k| = \begin{cases} a_0 & (k = 0) \\ \frac{1}{2}|g_k| & (k \neq 0) \end{cases}$$

$$\text{Arg}(C_k) = \begin{cases} +\phi_k & (k > 0) \\ -\phi_k & (k < 0) \end{cases}$$

Power Spectrum Two-Sided

$$\underline{|C_k|^2 + |C_{-k}|^2} = \frac{1}{2}g_k^2 = \frac{1}{2}(a_k^2 + b_k^2)$$

Periodogram One-Sided

$$2 \cdot |C_k| = \underline{g_k} = \underline{\sqrt{a_k^2 + b_k^2}}$$

CTFS and DTFS (1)

$$x(t) = \sum_{k=-\infty}^{+\infty} C_k e^{+jk\omega_0 t}$$

CTFS

$$C_k = \frac{1}{T} \int_0^T x(t) e^{-jk\omega_0 t} dt$$

$$k = \dots, -2, -1, 0, +1, +2, \dots$$

N Time Samples: N equations

$$t \rightarrow n T_s = n \left(\frac{T}{N} \right)$$

$$jk\omega_0 t \rightarrow k \left(\frac{2\pi}{T} \right) n \left(\frac{T}{N} \right) = \left(\frac{2\pi}{T} \right) nk$$

$$x(t) \approx \sum_{k=-M}^{+M} C_k e^{+jk\omega_0 t}$$

$N = 2M + 1$

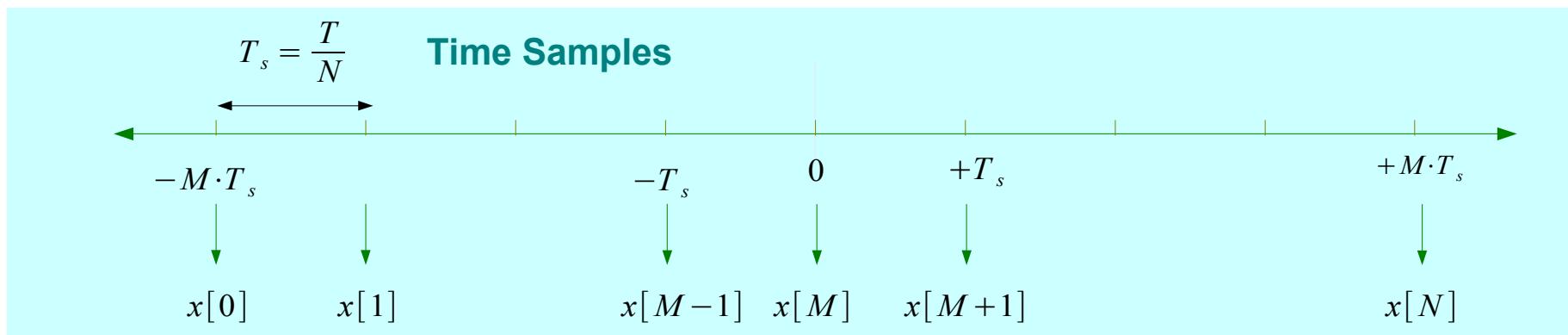
Truncate coefficients

$$k = -M, \dots, -1, 0, +1, \dots, +M$$

$$x[n] = \sum_{k=-M}^{+M} \gamma_k e^{+jk\omega_0 n \left(\frac{T}{N} \right)}$$

$$x[n] = \sum_{k=-M}^{+M} \gamma_k e^{+j \left(\frac{2\pi}{N} \right) nk}$$

$$n = 0, 1, 2, \dots, N-1,$$



CTFS and DTFS (2)

$$x(t) = \sum_{k=-\infty}^{+\infty} C_k e^{+jk\omega_0 t}$$

CTFS

$$C_k = \frac{1}{T} \int_0^T x(t) e^{-jk\omega_0 t} dt$$

$$k = \dots, -2, -1, 0, +1, +2, \dots$$

$$x[n] = \sum_{k=-M}^{+M} \gamma_k e^{+j\left(\frac{2\pi}{N}\right)nk}$$

$$n = 0, 1, 2, \dots, N-1,$$

$$\gamma_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\left(\frac{2\pi}{N}\right)nk}$$

$$k = -M, \dots, 0, \dots, +M$$

**Approximate Continuous Signal
With the truncated coefficients**

$$x_{FS}(t) = \sum_{k=-M}^{+M} \gamma_k e^{+jk\omega_0 t}$$

DTFS

$$C_k \approx \gamma_k = \frac{X[k]}{N}$$

Approximated Fourier Coefficients

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

DFT

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{+j(2\pi/N)kn}$$

IDFT

CTFS and DTFS (3)

Discrete Fourier Transform

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{+j(2\pi/N)kn}$$

$$x[n] = \sum_{k=-M}^{+M} \gamma_k e^{+j\left(\frac{2\pi}{N}\right)nk}$$

$$n = 0, 1, 2, \dots, N-1,$$

$$\gamma_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\left(\frac{2\pi}{N}\right)nk}$$

$$k = -M, \dots, 0, \dots, +M$$

$$x(t) = \sum_{k=-\infty}^{+\infty} C_k e^{+jk\omega_0 t}$$

CTFS

$$C_k \approx \gamma_k = \frac{X[k]}{N}$$

Approximated Fourier Coefficients

$$x_{FS}(t) = \sum_{k=-M}^{+M} \gamma_k e^{+jk\omega_0 t}$$

DTFS

 Truncate Fourier Coefficients

Power Spectrum using FFT

Discrete Fourier Transform

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{+j(2\pi/N)kn}$$

$$x(t) = \sum_{k=-\infty}^{+\infty} C_k e^{+jk\omega_0 t}$$

CTFS

$$x_{FS}(t) = \sum_{k=-M}^{+M} \gamma_k e^{+jk\omega_0 t}$$

DTFS

$$C_k \approx \gamma_k = \frac{X[k]}{N}$$

Approximated Fourier Coefficients

$$|C_k|^2 \approx \frac{|X[k]|^2}{N^2}$$

Approximated Power Spectrum

X = fft(x)

x = ifft(X)

Approximated Fourier Series Coefficients

fc = fft(x)/N = X/N

x = ifft(fc)*N

Periodogram using FFT

$$C_k \approx \gamma_k = \frac{X[k]}{N}$$

Approximated Fourier Coefficients

$$|C_k|^2 \approx \frac{|X[k]|^2}{N^2}$$

Approximated Power Spectrum

$$\frac{1}{N} \sum_{n=0}^{N-1} x^2[n] = \frac{1}{N^2} \sum_{k=0}^{N-1} |X[k]|^2$$

Average Power

→ $\left(\sqrt{\frac{\sum_{k=0}^{N-1} |X[k]|^2}{N}} \right)^2$

RMS of sq root Periodogram

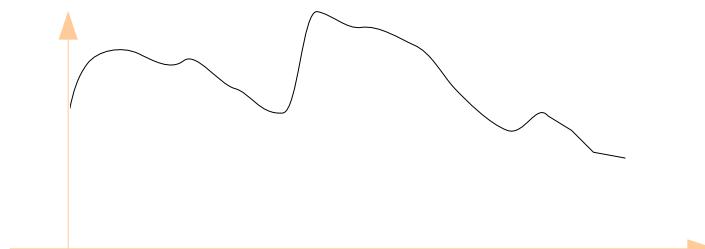
$$\frac{|X[k]|^2}{N} \quad k=0, 1, \dots, N-1$$

Approximated Periodogram

$$\frac{|X[k]|}{\sqrt{N}} \quad k=0, 1, \dots, N-1$$

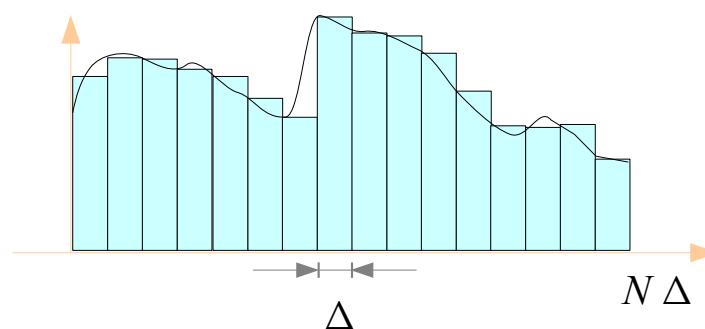
Square root Periodogram

RMS in continuous time



$$\frac{1}{T} \int_0^T g^2(t) dt$$

RMS in discrete time



$$\frac{1}{N\Delta} \sum_{k=0}^{N-1} |g[k]|^2 \Delta = \frac{1}{N} \sum_{k=0}^{N-1} |g[k]|^2$$

From CTFS to CTFT

Continuous Time Fourier Series

$$C_k = \frac{1}{T} \int_0^T x(t) e^{-jk\omega_0 t} dt \quad \leftrightarrow \quad x(t) = \sum_{n=0}^{\infty} C_k e^{+jk\omega_0 t}$$

$$C_k = \frac{1}{T_0} \int_{-T_0/2}^{+T_0/2} x_{T_0}(t) e^{-jk\omega_0 t} dt$$

$$C_k T_0 = \int_{-T_0/2}^{+T_0/2} x_{T_0}(t) e^{-jk\omega_0 t} dt$$

$$x_{T_0}(t) = \sum_{n=0}^{\infty} C_k e^{+jk\omega_0 t} \cdot \frac{2\pi}{2\pi} \cdot \frac{T_0}{T_0}$$

$$x_{T_0}(t) = \frac{1}{2\pi} \sum_{k=0}^{\infty} C_k T_0 e^{+jk\omega_0 t} \cdot \frac{2\pi}{T_0}$$

$$T_0 \rightarrow \infty \quad \omega_0 = \frac{2\pi}{T_0} \rightarrow d\omega \quad C_k T_0 \rightarrow X(j\omega) \quad x_{T_0} \rightarrow x(t)$$

Continuous Time Fourier Transform

$$X(j\omega) = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt \quad \leftrightarrow \quad x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{+j\omega t} d\omega$$

CTFS and CTFT

Discrete Fourier Transform

$$X[k] = \sum_{n=0}^{N-1} x[n] e^{-j(2\pi/N)kn}$$

$$x[n] = \frac{1}{N} \sum_{k=0}^{N-1} X[k] e^{+j(2\pi/N)kn}$$

$C_k \approx \gamma_k = \frac{X[k]}{N}$ Approximated Fourier Coefficients

$$x[n] = \sum_{k=-M}^{+M} \gamma_k e^{+j\left(\frac{2\pi}{N}\right)nk}$$

$$n = 0, 1, 2, \dots, N-1,$$

$$\gamma_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\left(\frac{2\pi}{N}\right)nk}$$

$$k = -M, \dots, 0, \dots, +M$$



Continuous Time Fourier Series

$$C_k T_0 = \int_{-T_0/2}^{+T_0/2} x_{T_0}(t) e^{-jk\omega_0 t} dt$$



$$T_0 \rightarrow \infty, \quad \omega_0 \rightarrow 0 \quad (\omega_0 \rightarrow d\omega)$$

$$x_{T_0}(t) = \frac{1}{2\pi} \sum_{k=0}^{\infty} C_k T_0 e^{+jk\omega_0 t} \cdot \frac{2\pi}{T_0}$$



Continuous Time Fourier Transform

$$X(j\omega) = \int_{-\infty}^{+\infty} x(t) e^{-j\omega t} dt$$



$$x(t) = \frac{1}{2\pi} \int_{-\infty}^{+\infty} X(j\omega) e^{+j\omega t} d\omega$$

FS Coefficients of Periodic and Aperiodic Signals

Frequency
Spacing

Periodic Signals

$$\Delta f = \frac{1}{N \Delta t}$$

Aperiodic Signals

$$\Delta f = \frac{1}{N \Delta t}$$

Two Sided
F.S. Coefficient

$$\frac{1}{N} X(k)$$

$$\frac{\Delta t}{N} X(k)$$

One Sided
F.S. Coefficient

$$\frac{1}{N} X(k) \quad k=0, \frac{N}{2}$$

$$\frac{\Delta t}{N} X(k) \quad k=0, \frac{N}{2}$$

$$\frac{2}{N} X(k) \quad k=1, \dots, \frac{N}{2}-1$$

$$\frac{2\Delta t}{N} X(k) \quad k=1, \dots, \frac{N}{2}-1$$

Frequency
Bin

$$k \Delta f$$

$$k \Delta f$$

Spectrum of Periodic Signals

Two-Sided Amplitude Spectrum

$$\begin{aligned} A_k &= \frac{1}{N}|X[k]| \\ &= \frac{1}{N}\sqrt{\Re^2\{X[k]\} + \Im^2\{X[k]\}} \\ k &= 0, 1, 2, \dots, N-1 \end{aligned}$$

One-Sided Amplitude Spectrum

$$\begin{aligned} \bar{A}_k &= \frac{1}{N}|X[0]| \quad k=0 \\ \bar{A}_k &= \frac{2}{N}|X[k]| \quad k=1, 2, \dots, N/2 \end{aligned}$$

Frequency Bin

$$f = \frac{k}{N} f_s$$

Phase Spectrum

$$\phi_k = \tan^{-1}\left(\frac{\Im\{X[k]\}}{\Re\{X[k]\}}\right) \quad k=0, 1, 2, \dots, N-1$$

Two-Sided Power Spectrum

$$\begin{aligned} P_k &= \frac{1}{N^2}|X[k]|^2 \\ &= \frac{1}{N^2}\left(\Re^2\{X[k]\} + \Im^2\{X[k]\}\right) \\ k &= 0, 1, 2, \dots, N-1 \end{aligned}$$

One-Sided Power Spectrum

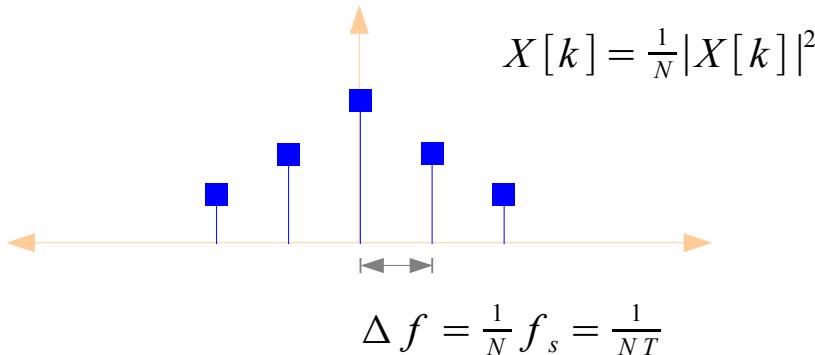
$$\begin{aligned} \bar{P}_k &= \frac{1}{N^2}|X[0]|^2 \quad k=0 \\ \bar{P}_k &= \frac{2}{N^2}|X[k]|^2 \quad k=1, 2, \dots, N/2 \end{aligned}$$

Frequency Bin

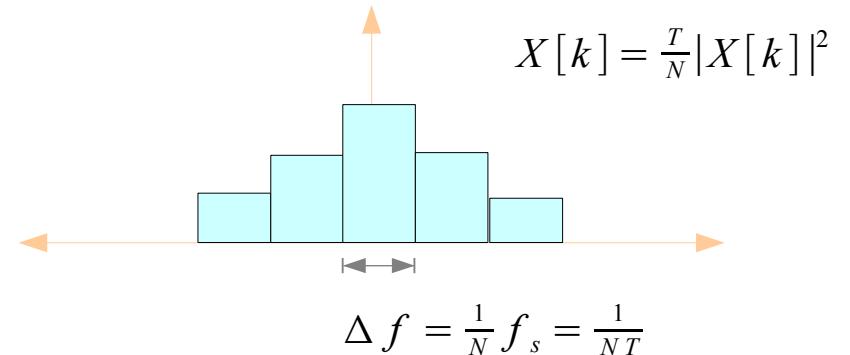
$$f = \frac{k}{N} f_s$$

Power Spectrum and Power Spectral Density

Power Spectrum



Power Spectral Density



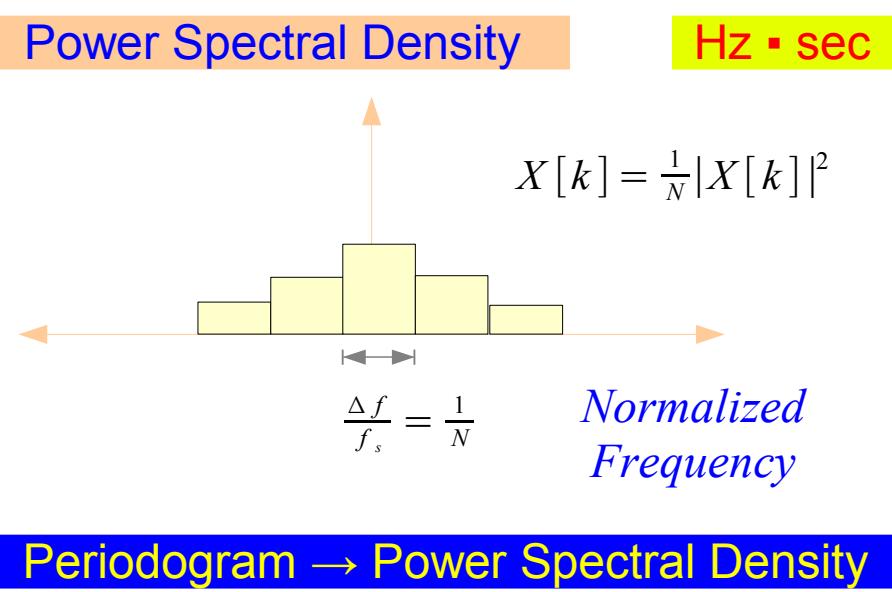
$$\frac{1}{N} \sum_{n=0}^{N-1} x^2[n] = \frac{1}{N^2} \sum_{k=0}^{N-1} |X[k]|^2$$

$$\Rightarrow \sum_{k=0}^{N-1} S[k] \Delta f$$

$$= \frac{1}{NT} \sum_{k=0}^{N-1} S[k] = \frac{1}{N^2} \sum_{k=0}^{N-1} |X[k]|^2$$

$$S[k] = \frac{T}{N} |X[k]|^2$$

Power Spectral Density



FS Coefficients of Random Signals

Random Signals

Frequency Spacing

$$\Delta f = \frac{1}{N \Delta t}$$

Two Sided
Power Spectral Density

$$P = \sum_{k=0}^{N-1} S(k) \Delta f$$

One Sided
Power Spectral Density

$$P = \sum_{k=0}^{N/2} S_1(k) \Delta f$$

$$S_1(k) = 2S(k) \quad k=0, \frac{N}{2}$$

$$S_1(k) = S(k) \quad k=1, \dots, \frac{N}{2}-1$$

$$\frac{1}{N \Delta t} \sum x^2 \Delta t$$

$$\sum S \Delta f = \frac{1}{N \Delta t} \sum S$$

$$S(k) = \frac{\Delta t}{N} |X(k)|^2$$

Frequency Bin

$$k \Delta f$$

Signals without discontinuity

Signals with discontinuity

Sampling frequency is not an integer multiple
of the FFT length

Leakage

$$[0, \frac{f_s}{2}]$$

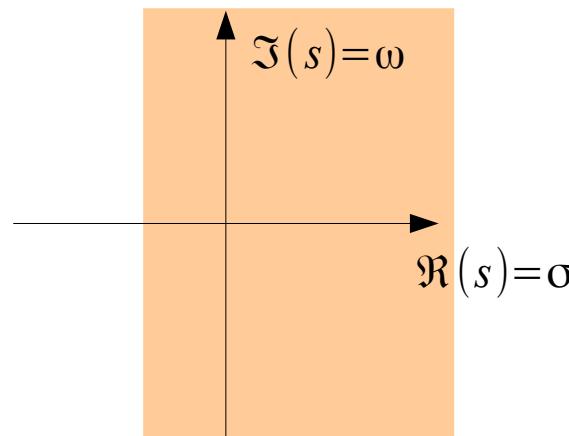
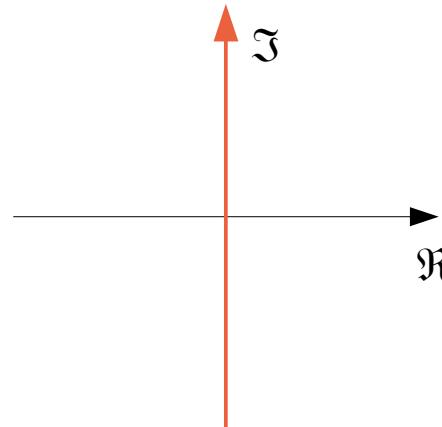
Fourier Transform

$f(t)$ A continuous sum of weighted exponential functions :

$$f(t) e^{-j\omega t}$$

$$-\infty < \omega < +\infty$$

Not so useful in transient analysis



Laplace Transform

$$f(t) e^{-st} = f(t) e^{-(\sigma + j\omega)t}$$

Linear Time Domain Analysis

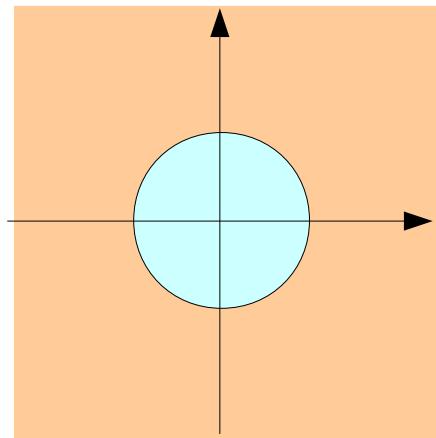
Initial Condition

z Transform

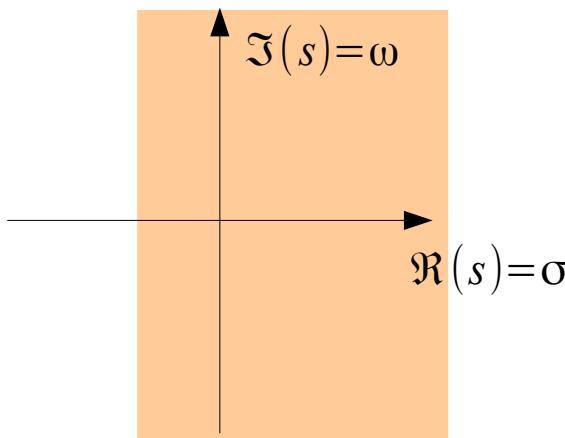
$$f[n] z^{-n}$$

Discrete Time System

Difference Equation



$$z = e^{sT} = e^{\sigma T} e^{j\omega T}$$



References

- [1] <http://en.wikipedia.org/>
- [2] J.H. McClellan, et al., Signal Processing First, Pearson Prentice Hall, 2003
- [3] A “graphical interpretation” of the DFT and FFT, by Steve Mann