# Formatting (2A)

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### Formatting and Source Coding

#### **Formatting**

Make the source signal compatible with digital processing

#### **Transmit Formatting**

A transformation from source information to digital symbols

### **Source Coding**

Formatting + Data Compression

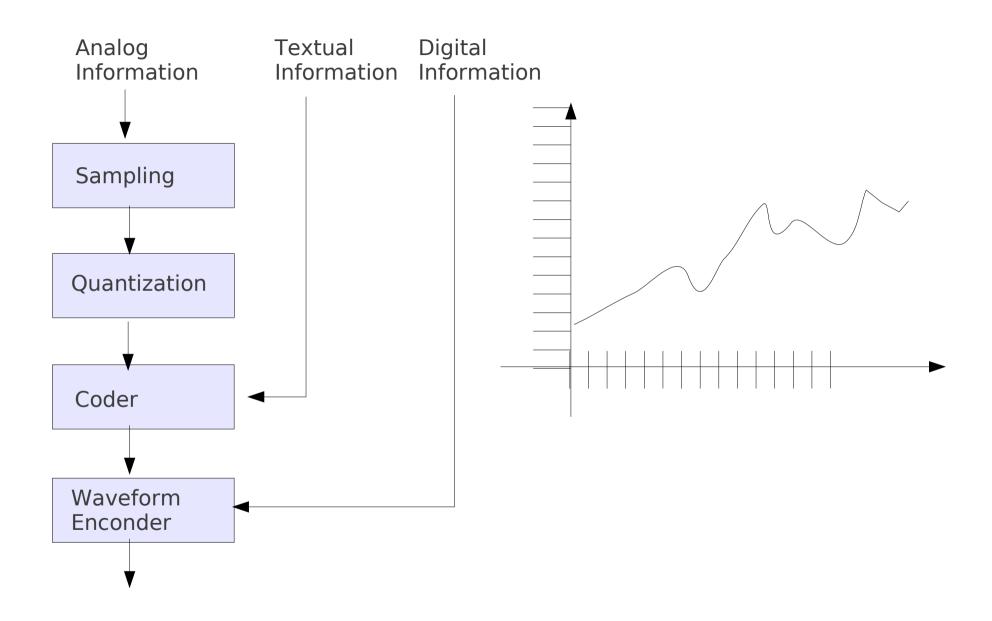
### **Baseband Signal**

From DC up to some finite frequency (< a few MHz)
Transmitted over the cable
Not appropriate to transmit over long distance → Bandpass Mod

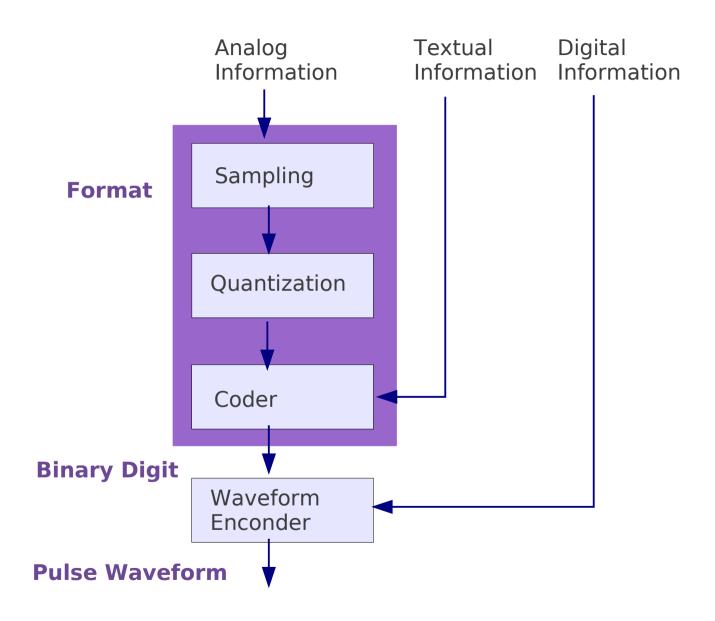
#### **Pulse (Baseband) Modulation**

Pulse waveforms are assigned that represent formatted symbols

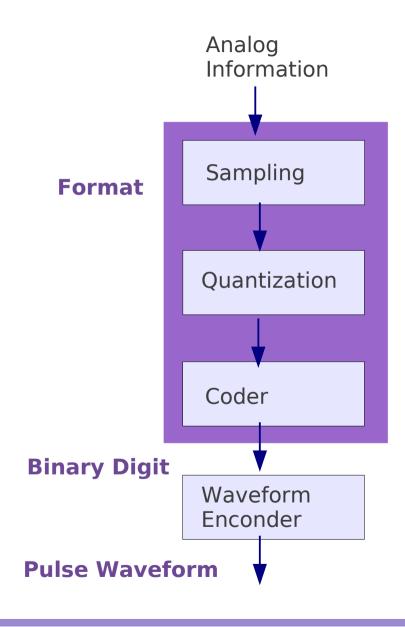
# Energy and Power Spectral Densities (2)

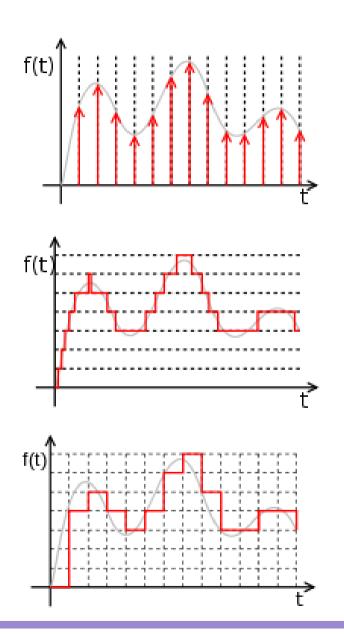


### **Baseband Signal**

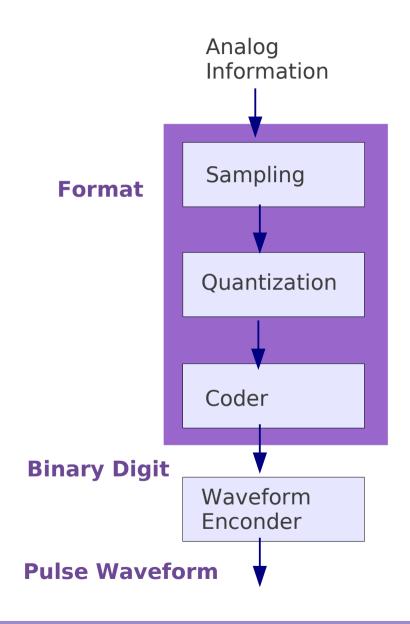


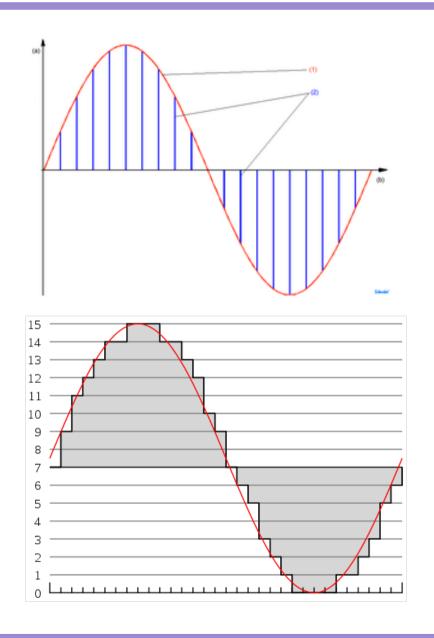
# Sampling and Quantization





### PAM and PCM





### 8-ary Symbol

T H I N K Message  $001010 \ 000100 \ 100100 \ 011100 \ 110100$  6-bit ASCII  $001 \ 010 \ 000 \ 100 \ 100 \ 100 \ 011 \ 100 \ 110 \ 100$  8-ary digits '(symbols)  $s_1(t) \ s_2(t) \ s_0(t) \ s_4(t) \ s_4(t) \ s_3(t) \ s_4(t) \ s_6(t) \ s_4(t)$  8-ary waveform

## **Binary Symbol**

T H I N K Message

001010 000100 100100 011100 110100 6-bit ASCII

 $0\ 0\ 1\ 0\ 1\ 0\ 0\ 0\ 1\ 0\ 0\ 1\ 0\ 0\ 0\ 1\ 1\ 1\ 0\ 0\ 1\ 1\ 0\ 0$ 

001010 00010010010001110011, binary digits (symbols)

S<sub>0</sub>(t) S<sub>1</sub>(t) S<sub>2</sub>(t) S<sub>3</sub>(t) S<sub>3</sub>(t) S<sub>3</sub>(t) S<sub>3</sub>(t) S<sub>3</sub>(t)

binary waveform

### Impulse Sampling

#### Impulse train

$$X_{\delta}(t) = \sum_{n=-\infty}^{+\infty} \delta(t - nT_s) \qquad \longleftrightarrow \qquad X_{\delta}(f) = \frac{1}{T_s} \sum_{n=-\infty}^{+\infty} \delta(f - nf_s)$$

Shifting property

$$x(t)\delta(t-t_0) = x(t_0)\delta(t-t_0)$$

$$\begin{array}{lll} x_s(t) &=& x(t)x_\delta(t) & & \longleftarrow & X_s(f) &=& X(f) * X_\delta(f) \\ &=& \sum\limits_{n=-\infty}^{+\infty} x(t)\delta(t-nT_s) & &=& X(f) * \left[\frac{1}{T_s}\sum\limits_{n=-\infty}^{+\infty} \delta(f-nf_s)\right] \\ &=& \sum\limits_{n=-\infty}^{+\infty} x(nT_s)\delta(t-nT_s) & &=& \frac{1}{T_s}\sum\limits_{n=-\infty}^{+\infty} X(f-nf_s) \end{array}$$

### **Natural Sampling**

#### Pulse train

$$x_p(t) = \sum_{n=-\infty}^{+\infty} c_n e^{j2\pi n f_s t} \qquad \qquad \qquad \qquad \qquad \qquad \qquad \qquad c_n = \frac{1}{T_s} sinc(\frac{nT}{T_s})$$

$$\begin{array}{lll} x_s(t) &=& x(t)x_p(t) & & & \longrightarrow & X_s(f) &=& X(f) \, * \, X_p(f) \\ \\ &=& x(t)\sum_{n=-\infty}^{+\infty} c_n e^{j2\pi f_s t} & & & \\ \\ &=& \sum_{n=-\infty}^{+\infty} c_n \big[x(t)e^{j2\pi f_s t}\big] & & \longrightarrow & = \sum_{n=-\infty}^{+\infty} c_n X(f-nf_s) \end{array}$$

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### Sample and Hold

#### Sampled Pulse train

$$x_p(t) = \sum_{n=-\infty}^{+\infty} c_n e^{j2\pi n f_s t}$$



$$c_n = \frac{1}{T_s} sinc(\frac{nT}{T_s})$$

$$x_s(t) = p(t)*[x(t)x_{\delta}(t)]$$



$$X_s(t) = p(t)*[x(t)x_s(t)]$$
  $\longrightarrow$   $X_s(f) = X(f) * X_p(f)$ 

$$= p(t) * \left[ x(t) \sum_{n=-\infty}^{+\infty} \delta(t-nT_s) \right]$$



$$= p(t) * \left[ x(t) \sum_{n=-\infty}^{+\infty} \delta(t-nT_s) \right] \qquad = P(f) \left[ X(f) * \left[ \frac{1}{T_s} \sum_{n=-\infty}^{+\infty} \delta(f-nf_s) \right] \right]$$

$$= P(f) \left[ \frac{1}{T_s} \sum_{n=-\infty}^{+\infty} X(f - n f_s) \right]$$

### Sampling Theorem

#### **Uniform Sampling Theorem**

A bandlimited signal having no spectral components above  $\mathbf{f_m}$  Hz can be determined uniquely by values sampled at *uniform intervals* of  $\mathbf{T_c}$  seconds

$$T_s \leq \frac{1}{2f_m} \qquad f_s = \frac{1}{T_s}$$

Upper limit of T

$$f_s \ge 2f_m$$

Lower limit of **f**<sub>s</sub>

**Nyquist Criterion** 

Nyquist Rate  $f_s = 2 f_m$ 



# Ensemble Average





# Time Averaging and Ergodicity



# Time Averaging and Ergodicity

#### References

- [1] http://en.wikipedia.org/
- [2] http://planetmath.org/
- [3] B. Sklar, "Digital Communications: Fundamentals and Applications"